Towards Mobile Data Centres: Provision of End-to-End 10 and 40 Gbit/s Ethernet Train Backbones on International Rolling Stock

Felix Ngobigha, Geza Koczian, and Stuart D. Walker

Abstract—The next generation of railway customer-oriented services are expected to generate a large volume of data (≈ 10s of TB). As a result, passengers’ applications, safety, security, and Internet-on-Board (IoB) sensors challenge current Train Communication Networks. With the present Ethernet Train Backbone (ETB) specification of just 100 Mbit/s in total, railway passenger services will not support intelligent, seamlessly connected and mobile media on-board trains. In this paper, we propose a novel ETB design with experimental results demonstrating end-to-end 10 Gbit/s and 40 Gbit/s throughput results over existing conducting media on commercial railway carriages. This is equivalent to 100 Mbit/s per user on real-world railway rolling stock and shows that standard RailCat 5e cabling and new rail-approved 10 Gbit/s ETB active nodes (switches) fully support emerging trends and future-proof ETB configurations.

Index Terms—ETB train regions, passive components, active network devices, Ethernet consist network, twisted-quad cable, quadrupole.

I. INTRODUCTION

Eff ective communications have always been identified as one of the essential operational parameters for railway operation. Historically, there have been challenges in terms of non-interoperability. As a result, a set of international standards such as IEC 61375 were developed [1], [2], to overcome this situation namely: the Train Communication Network (TCN) which defines a system architecture and the necessary protocols for communication at train and vehicle level [3] was first introduced. It consists of a Multifunction Vehicle Bus (MVB) [4] inside each carriage and the Wire Train Bus (WTB) [5] to connect the different vehicles. The TCN components have been standardized in IEC 61375 as above. The current TCN configuration comprises the Ethernet Train Backbone (ETB) [6] and Ethernet Consist Network (ECN) [7] with a specification up to 100 Mbit/s with physical links along the train to connect the active network devices together. This is perceived as inadequate, for example, a 500 passenger train would offer just 200 kbit/s per passenger. Providing a rich movie experience such as H.265 real-time 4K video streaming at 15 Mbit/s for 500 passengers requires 7.5 Gbit/s. Even higher loads are possible noting that 130 - 180 passengers for 9 carriages during peak periods [8], [9] will consume 24.3 Gbit/s. This introduces a massive challenge to the train operating companies (TOCs) and forms the motivational basis for this work. Without the ultra-high capacity ETB described here, advances in train-to-shore 5G wireless (for example) and beyond will be neutralized [10]–[12]. Various technologies for inter-carriage connection such as millimetre wave (mmWave) radio, optical wireless, and fibre-optic communication [13]–[18] have all been considered for ETB applications. However, these have performed badly in the harsh railway environment, for example: wide temperature range, humidity, electromagnetic interference, water ingress, abrasive substances, and large-scale mechanical misalignment due to carriage sway and track curvature.

Prior to this study, ETB realization used standard conducting media technologies which were limited to 100 Mbit/s maximum due to inappropriate cable and connector provision. These fell far short of the seamless higher data rate connectivity required for on-board operation-related-services and active users. Due to environmental barriers and sensitivity to blockages and shadowing at higher frequencies as with mmWwave, the radio link between carriages must be carefully designed to counter non-line-of-sight (NLOS) situations, typically by using beam forming techniques [19]. Using the subset of optical wireless communications technologies [20] such as visible light communication (VLC) and free space optics (FSO) for ETB has also shown susceptibility to atmospheric conditions such as fog and dust. Additionally, the unquestioned capacity of fibre-optic cables in communication systems together with low power loss, and immunity to electromagnetic interference has to be weighed against the harsh abrasive railway environment.

A discussion around future railway services and bandwidth requirements was presented [14] based on lower frequency wireless technologies, including inter-carriage connection based on IEEE802.11 [21], WiMAX [22], [23] with dedicated short-range communications [18]. Guan et al [14] also pointed out the difficulty of accurately analyzing the on-board bandwidth with respect to the number of active users during peak period, simultaneous communication within certain scenarios, and different service types. In summary, we show here that the provision of end-to-end, future-proof ETB
In real-world applications have proven to be successful in terms of seamlessly supporting ETB physical layer link speed of 10 Gbit/s. The receptacle is a contact device installed at the outlet for the connection of an attached plug (known as a Jumper in rolling stock industry parlance). The rail-approved connector design does not adequately account for electromagnetic shielding and crosstalk between neighbouring pairs of the four pins; such crosstalk being at variance with IEEE 802.3an (10GBASE-T) standards [25]. Based on the intuitive analysis of the pins connector in terms of electromagnetic interference and crosstalk, it became apparent that the configuration is similar to an electromagnetic non-collinear dipole pair (quadrupole) as discussed [26]. The requisite arrangement was initially applied to a section (2 x 4-wire RailCat 5e twisted-quad cables) and terminated onto 2 x rail-approved connectors by employing this quadrupole configuration based on balanced/symmetrical fields on the four pins termination. This minimizes the effect of crosstalk between cable quads with equal and opposite fields having a cancellation effect. Alternatively, since the next improvement is expected to be four times higher in throughput, the configuration was changed slightly in contrast to the approach described in [24] by utilizing an arrangement of 8 x 4-wire RailCat 5e twisted-quad copper cables as the inter-consist or inter-car Jumper system with careful consideration to reducing the effect of alien crosstalk (which occurs when cable bundles run in close proximity). The RailCat cables cross-sections used for this experiment are shown in Figures 2 and 3.

II. ETB PHYSICAL LAYER MODEL

In this section, the design principles of the proposed ETB physical layer model are presented based on [24] and determine the feasibility of utilizing passive guided media methods to achieve higher data rates in retrofit and future railway passenger-oriented services. The main requirements are to minimise crosstalk/electromagnetic interference within the connectors, reduce termination issues for RailCat 5e cable, and avoid using 4 x 8-wire braided Shielded Foil Twisted-Pair (S/FTP) cabling for the ETB physical layer of the train regions such as inter-car and inter-consist.

The schematic of the ETB configuration design with corresponding standard RailCat 5e twisted-quads for the Inter-car and/or Inter-consist with the connectors and RailCat 7 within the rolling stock as shown in Figure 1. This assembly
TABLE I
COMPARISON OF TWISTED COPPER CABLE BASED ETHERNET PHYSICAL
(PHY) MEDIUM CHARACTERISTICS

<table>
<thead>
<tr>
<th>Standard</th>
<th>Nominal Speed (Mbit/s)</th>
<th>Pairs reqd.</th>
<th>Cable requirement</th>
<th>Max. Cable length (m)</th>
<th>Bandwidth (MHz)</th>
</tr>
</thead>
<tbody>
<tr>
<td>802.3a</td>
<td>100</td>
<td>2</td>
<td>Cat 5e</td>
<td>100</td>
<td>100</td>
</tr>
<tr>
<td>802.3ab</td>
<td>1,000</td>
<td>4</td>
<td>Cat 5e</td>
<td>100</td>
<td>100</td>
</tr>
<tr>
<td>802.3an</td>
<td>10,000</td>
<td>4</td>
<td>Cat 6A</td>
<td>100</td>
<td>500</td>
</tr>
<tr>
<td>ISO/IEC 11801</td>
<td>10,000</td>
<td>4</td>
<td>Cat 7</td>
<td>100</td>
<td>600</td>
</tr>
</tbody>
</table>

TABLE II
DYNAMIC (FLEXING) APPLICATION TEST RESULTS OF RAIL-APPROVED
CABLES FOR THE IMPLEMENTATION OF THE INTER-CONSIST OR
INTER-CAR CABLING

<table>
<thead>
<tr>
<th>Cable</th>
<th>Cable Construction</th>
<th>Cycles to Failure (CF)</th>
<th>CFdiff(%)</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td>minCF</td>
<td>maxCF</td>
</tr>
<tr>
<td>Cat 6A</td>
<td>4 x [2 X 0.20 mm² (24 AWG)]</td>
<td>3,250</td>
<td>30,000</td>
</tr>
<tr>
<td>Cat 7</td>
<td>4 x [2 X 0.20 mm² (24 AWG)]</td>
<td>1,330</td>
<td>27,895</td>
</tr>
<tr>
<td>Cat 5e</td>
<td>4 x 0.33 mm² (22 AWG)</td>
<td>27,500</td>
<td>66,700</td>
</tr>
<tr>
<td>Cat 5e</td>
<td>4 x 0.5 mm² (20 AWG)</td>
<td>88,000</td>
<td>132,500</td>
</tr>
</tbody>
</table>

Considering the results from the dynamic test conducted with different cable construction as shown in Table II, we evaluated the cycles to failure difference as:

\[
CF_{diff}(\%) = \left( \frac{\max_{CF} - \min_{CF}}{\max_{CF} + \min_{CF}} \right) \times 100.
\] (1)

The flexing test results tabulated in Table II shows the limit for performance and reliability to evaluate the mechanical fracturing properties of the sample RailCAT cables. Likewise, the empirical rule [29] was used to determine whether a given data set is normally distributed and to find outliers, which may be the result of experimental errors. As shown in Table II, the RailCat 5e twisted-quad construction with the 20 AWG is preferred for the intended applications of inter-car and/or inter-consist in the ETB train region and has a larger cycles to failure. While there is standard for the RailCat cabling, ETB face unique challenges when managing electrical properties and other factors. Without proper planning, applications could experience many issues, including disrupted network communication and/or inaccurate performance. Therefore, it is necessary to understand the implementation requirements needed for a given application to select a suitable cable for the ETB inter-car and inter-consist region. Furthermore, the flexing test cycling rate for this work was also increased from 15 cycles per minute as implemented in [24] to 20 cycles per minute, this value is considered as a worst case scenario in terms of the carriage sway at maximum speed ≥ 250 km/h, achieved by Eurostar trains on the HS1 line between London and the Channel Tunnel [30].

A. Quadrupole analysis

In accomplishing minimal interference within the pins connector, it emerged that the optimum configuration is similar to an electromagnetic quadrupole [26], [31]. We consider a finite length of 4-wire RailCat 5e twisted-quad copper cabling for the ETB train region of inter-car and inter-consist, stripping back considerable length of conductors and terminated onto the connector. This approach will be very relevant in the implementation of the electromagnetic quadrupole and has been discussed in transmission quadrupole mass spectrometer (filter) which historically proved to be very successful in detecting only ions of a certain mass-to-charge ratio particles [32]–[34]. In [26], we noticed that, if all the charges are distributed along an axis (the z-axis for example in a linear quadrupole), then \( \cos ix = \cos jy = 0 \), and \( \cos kz = 1 \). In the case of non-collinear discrete charge points, the symmetry axes are as shown in Figure 4. It is assumed that the transmitting pair denotes +ve charges while the receiving pair represent −ve charges for the quadrupole analysis.
The analysis started by defining the angles in Figure 4 relative to point \( P(r, \theta, 0) \):

\[
\begin{align*}
    a &= \theta; \\
    b &= \frac{\pi}{2} + \theta \\
    c &= \pi - \theta; \\
    d &= \frac{\pi}{2} - \theta,
\end{align*}
\]

where \( P \) is an arbitrary field point. Considering Coulomb’s Law, if the localized charge distribution has a net electric charge \( Q_{\text{net}} \) at far distance, then potential \( V(\bar{r}) \) to a good approximation will behave very much like that of a point charge:

\[
V_{\text{far}}(\bar{r}) = \frac{1}{4\pi \varepsilon_0} \frac{Q_{\text{net}}}{r}
\]

and

\[
\vec{E}_{\text{far}}(\bar{r}) = -\nabla V_{\text{far}}(\bar{r}) = -\frac{1}{4\pi \varepsilon_0} \frac{Q_{\text{net}}}{r^2}
\]

where \( \frac{1}{4\pi \varepsilon_0} = K \) is the Coulomb constant. Using the Law of Cosines [35]:

\[
c^2 = a^2 + b^2 - 2ab \cos \theta,
\]

the source charge locations for the charges with respect to \( P \) are given as:

\[
r_a^2 = \left( \frac{d}{2} \right)^2 + r^2 - 2 \left( \frac{d}{2} \right) r \cos \pi
= r^2 + \left( \frac{d}{2} \right)^2 - dr \cos \theta,
\]

\[
r_b^2 = \left( \frac{d}{2} \right)^2 + r^2 - 2 \left( \frac{d}{2} \right) r \cos \left( \frac{\pi}{2} + \theta \right)
= r^2 + \left( \frac{d}{2} \right)^2 + dr \sin \theta
\]

Inferring to a point charge of equation (3), the total potential of the configuration can be evaluated using the principle of superposition:

\[
V_{\text{quad}}(\bar{r}) = V_{+r_a}(\bar{r}) + V_{-r_b}(\bar{r}) + V_{+r_c}(\bar{r}) + V_{-r_d}(\bar{r})
\]

substituting equations (6) - (9) into equation (10) gives:

\[
V_{\text{quad}}(\bar{r}) = \frac{Q K}{r \sqrt{1 + \left( \frac{1}{4} \right) \left( \frac{d}{r} \right)^2 - \left( \frac{d}{r} \right) \cos \theta}}
- \frac{Q K}{r \sqrt{1 + \left( \frac{1}{4} \right) \left( \frac{d}{r} \right)^2 + \left( \frac{d}{r} \right) \sin \theta}}
+ \frac{Q K}{r \sqrt{1 + \left( \frac{1}{4} \right) \left( \frac{d}{r} \right)^2 - \left( \frac{d}{r} \right) \cos \theta}}
- \frac{Q K}{r \sqrt{1 + \left( \frac{1}{4} \right) \left( \frac{d}{r} \right)^2 + \left( \frac{d}{r} \right) \sin \theta}}
\]

It is worth noting that this is not an approximate mathemati- cal expression for potential associated with pure physical non-colinear quadrupole with charges separated from each other by an equal distance \( d \) along the \( y-z \) plane. Careful choice of coordinate system simplified the mathematics and clarified the physical foundations of the configuration. We reiterate, briefly the development of the electric potential and field intensity for the quadrupole as set out in [26] and [31]:

\[
V_{\text{quad}} = \frac{1}{4\pi \varepsilon_0} \left( \frac{1}{|\bar{r}|^3} \right) \sum_{ij} \frac{1}{2} Q_{ij} n_i n_j
\]
the components of the $Q$ matrix are defined by:

$$Q_{ij} = \sum_l Q_l \left(3r_{il}r_{jl} - |\vec{r}_l|^2 \delta_{ij}\right), \quad (13)$$

where the indices $i, j$ run over the $y-z$ plane and $\delta_{ij}$ is the Kronecker delta [36].

As with any multipole moment, if a lower-order, monopole or dipole, in this case, is non-zero, then the value of the quadrupole moment depends on the choice of the coordinate origin. In our approach to terminating the connectors as shown in Figure 4, we expected that the quadrupole moment will be reduced to zero and coordinate dependent. The associated electric fields can be evaluated using the potential field approach [37] and [38]:

$$E_{quad} = -\nabla V_{quad}. \quad (14)$$

The negative sign in equation (14) is arbitrary but has been chosen to conform with the convention which directs the vector $E_{quad}$ outward from a positive discrete charge, and in spherical coordinates form:

$$E_{quad} = E_r + E_\theta + E_\phi = -\nabla V_{quad}. \quad (15)$$

Explicitly writing out the form of the electric field intensity $E_{quad}$ for an electric quadrupole with coordinate dependent gives:

$$E_{quad} = \frac{3 \cdot 2Qd^2}{4\pi\varepsilon_0} \left(\frac{1}{r^4}\right) \times \left[\frac{3\cos^2\theta - 1}{2}\right] \hat{r} + \sin \theta \cos \theta \hat{\theta}, \quad (16)$$

where

$$\left(\frac{3\cos^2\theta - 1}{2}\right) = P_n(x)$$

are legendre polynomials of degree 2 with no $\phi$-dependence because the charge configuration is manifestly axially symmetrical in the $y-z$ plane (invariant under arbitrary $\phi$ rotation). If we assumed that the value of $\theta = 0$, then (16) becomes

$$E_{quad} = \frac{3 \cdot 2Qd^2}{4\pi\varepsilon_0} \left(\frac{1}{r^4}\right) \hat{r}. \quad (17)$$

With this analysis, the anticipated bottleneck as a result of the crosstalk within cable and the electromagnetic interference from cables in close proximity has been minimised with improved system performance and higher throughput achieved for the intended applications.

**B. Simulation of quadrupole potential and electric field**

Other than the quadrupole theoretical analysis of the selected components such as connectors and cables dedicated to the Ethernet. We simulate the quadrupole potential and electric field intensity by assuming four discrete charges, equal magnitude, alternating signs at some distance as shown in Figure 4 which is similar to the configuration described [24]. Evaluation of equation (12) is carried out by adapting the MATLAB code designed for the simulation of a pure, linear (i.e. axially/azimuthally symmetric) electric quadrupole with the moment oriented along the z-axis and using MATLAB version 9.6 R2019a on a 64-bit machine. Figures 5 to 7 show an example of a RailCat 5e cable terminated onto a connector to evaluate the electric potential in three and two-dimensions corresponding to the arrangement of Figure 4.

**Fig. 5.** 3D quadrupole potential of four charges, equal magnitude and alternating signs

**Fig. 6.** 2D quadrupole potential of four charges, equal magnitude and alternating signs

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Numerical analysis of the quadrupole electric field intensity of equation (17) is also undertaken following the above approximation where the RailCat 5e cable is terminated onto a connector is shown in Figure 8 also suggests similar patterns is achieved.

![Fig. 7. 2D quadrupole potential of four charges, equal magnitude and alternating signs with field lines](image)

III. NUMERICAL RESULTS AND DISCUSSION

A. Field tester summary results

The summary field tester results using the Fluke Networks CableAnalyzer DSX-8000 for each section of the configuration followed a similar pattern. Satisfactory results were obtained for Transverse Conversion Loss (TCL) which validates how balanced the termination with respect to the RailCat 5e cable and the enhanced connector. Near End Cross Talk (NEXT), Return Loss and Insertion Loss headroom worst-case margins on both the main and smart remote units of the Fluke tester were satisfactory. Obviously, the results were well above the minimum requirements as specified in IEEE 802.3an (10GBASE-T) standards for each section and demonstrated the robustness of our new approach to provide the needed bandwidth at 10 and 40 Gbit/s ETB physical layer speed. The Telecommunications Industry Association (TIA) test limits used in the validation process are more stringent compared with the IEEE 802.an specification.

![Fig. 8. Quadrupole electric field intensity of four charges, equal magnitude and alternating signs with directions](image)

![Fig. 9. NEXT of the complete ETB configuration](image)

<table>
<thead>
<tr>
<th>Test limit</th>
<th>TCL (dB)</th>
<th>NEXT (dB)</th>
<th>Return Loss (dB)</th>
<th>Insertion Loss (dB)</th>
<th>Test status</th>
</tr>
</thead>
<tbody>
<tr>
<td>TIA Cat 6A</td>
<td>41.6</td>
<td>8.2</td>
<td>10.6</td>
<td>45.0</td>
<td>Pass</td>
</tr>
<tr>
<td>TIA Cat 6A</td>
<td>41.8</td>
<td>10.0</td>
<td>11.3</td>
<td>44.8</td>
<td>Pass</td>
</tr>
<tr>
<td>TIA Cat 6A</td>
<td>40.6</td>
<td>5.4</td>
<td>10.7</td>
<td>44.9</td>
<td>Pass</td>
</tr>
<tr>
<td>TIA Cat 6A</td>
<td>40.9</td>
<td>8.1</td>
<td>9.0</td>
<td>44.7</td>
<td>Pass</td>
</tr>
</tbody>
</table>

The numerical results tabulated in Table III for the TCL, NEXT, and Return Loss are shown in terms of worst case margins respectively. While the Insertion Loss values are evaluated with respect to loss margin as tabulated in Table III. The experimental results presented in Table III considering the TIA Channel Cat 6A test limits are in good agreement and follow similar pattern as given in [24]. Figures 9 - 11 are extracted graphs which suggest that the new ETB configuration clearly exceeds the threshold in accordance with TIA Channel 6A test limits.

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Return Loss (RL) is a measure of all reflections that are caused by the impedance mismatches at all locations along the ETB physical layer region. Obviously, a good match between characteristic impedance and termination resistance in the nodes provides for a good transfer of power/data to and from the link and minimize reflections. The return loss measurement varies significantly with frequency. The grey section of the limit line as shown in Figure 10 indicates that the insertion loss (IL) \(< 3\) dB and not evaluated against the test limit (the 3 dB Rule means the integrity of the signal will not be affected below the noise floor or test limit). Sources of return loss are mostly due to variations in the value of the characteristic impedance along the cable and mainly from connectors. Considering the measurements carried out with the field tester, we feel confident that this configuration will perform well, even in a noisy environment such as the harsh railway environment.

Note further that, the test limit line shown in Figures 9 has a low-frequency flat region at 65 dB in the NEXT is often referred to as the measurement plateau. This allows for dynamic range in between the measurement noise floor of the test equipment (either Vector Network Analyzer or field tester) and the limit line. It should be noted that without the plateau, the limit becomes more challenging at low frequency and would cause worse failure. This low-frequency flat region limit line is also applicable to other parameters such as Return Loss at 19 dB and Transverse Conversion Loss at 40 dB and are specified by ISO/IEC JTC1 SC25 WG3 [39] and the TIA TR42.7 committees which form part of the required compliance testing for a given category/class. In the case of the TIA Cat 6A channel test limit, the flat region at the low frequency for the NEXT test limit line is less stringent in comparison to the ISO11801 Channel Class Fd and F test limits.

Figure 11 gives an indication of balance measurement with clear margin between the data points and the test limit to ensure that any noise injected into the cabling is cancelled out. This is accomplished by injecting a differential mode signal into each configuration pair, then measuring the common mode signal returned on that same cabling with a high TCL value means that the impedances of the conductors relative to ground are almost equal. High TCL values correspond to better noise immunity and lower emissions in terms of electromagnetic interference from other sources considering the harsh railway environment. The TCL was not considered as an issue in our previous work [24]. However, the suitability of running the ETB in close proximity with other signal cables was investigated through experimental setups. Generally, the installation of data cabling infrastructure such as twisted-pair copper assembly in close proximity to electrical power cables is anticipated to be susceptible to electromagnetic interference due to harmonics. The resulting effect could adversely disrupt the signal and the communication would become noisy and garbled. In many cases, transmissions will simply not make it. In other cases, transmission rates will slow down as communications are constantly retried. In the event that the data cabling must go near electrical power lines, the norm demands to cross them in perpendicular which was the approach adapted in §II-A with the termination of the improved connectors (e.g., proper shielding to prevent electromagnetic fields from entering or leaving the configuration).

B. ETB experimental results

To validate the ruggedness of our new experimental approach to seamlessly support a 40 Gbit/s link speed over a distance of 55-metres (i.e. length of a passenger car is taken to be 25-metre, a total of 50-metre for the two cars and the inter-car and/or inter-consist is considered to be 5-metre), we use an open-source software known as the Microsoft Network test transmission control protocol (MS NTtcp) [40]. This sends and/or receives a randomly generated data stream from the ETB active nodes through the 40 Gbit/s servers from in-memory buffers, so there is no disk input/output to limit its throughput. This feature makes it a very useful tool for...
measuring network performance and throughput analysis. The input parameters are systematically chosen within an allowed set such as Ethernet frame sizes, and optimization of the MS NTtcp software variables to fit the intended application.

As shown in Table IV, the NTtcp optimized variables used for the experimental setup are defined as follows: -l specifies the length of each buffer that NTtcp uses to transfer data, -rb represents the total per-socket buffer space reserved and in turn, controls the receivers transmission control protocol (TCP) window size, -n specifies the number of buffers that NTtcp transfers before it stops and reports the results, -a enables asynchronous data transfer mode, -w directs NTtcp to use Win32 (WSASend and WSAREcv) functions to transfer data, -v enables verbose mode, and -fr specifies full receive buffer mode.

| TABLE IV | MICROSOFT NTTCP SOFTWARE INPUT PARAMETERS SETTINGS FOR THE TESTED |
|-----------------|-----------------|-----------------|-----------------|
| Transmitting server/switch setting | Receiving server/switch setting | |
| -l 1048576 | -l 1048576 | 
| -rb N/A | -rb 220000 |
| -n 100000 | -n 100000 |
| -a 16 | -a 16 |
| -w No value | -w No value |
| -v No value | -v No value |
| -fr No value | -fr No value |

Taking into consideration the maximum transmission unit (jumbo frame) [41] and [42] used in this experimental setup, the overall output results are shown in Table V using the NTtcp software to realise end-to-end 40 Gbit/s throughput clearly demonstrates the robustness and reliability of our new approach over standard RailCat cabling and enhanced connector with minimum retransmission or no errors. This suggests that the link speed is approximately 40 Gbit/s over four 10 Gbit/s links. The number of packets sent and packets received in bytes, as shown in Table V for both outputs are in good correlation. The retransmission rate is calculated either in terms of segments or bytes but in this study, the latter was chosen and the byte retransmission rate was evaluated to be less than 1.0 % in accordance with [43].

| TABLE V | SUMMARY TEST RESULTS OF USING MICROSOFT NTTCP BENCHMARK SOFTWARE TO VALIDATE 40 GBIT/S THROUGHPUT OVER THE RAILCAT CABLING/ENHANCED CONNECTORS CONFIGURATION |
|-----------------|-----------------|-----------------|-----------------|
| Tx. outputs | Rx outputs | |
| Total Bytes (Mega) | 415146.97 | Total Bytes (Mega) | 415197.30 |
| Realtime (s) | 84.69 | Realtime (s) | 86.70 |
| Frame Size (Bytes) | 9000 | Frame Size (Bytes) | 9000 |
| Total Throughput (Mbit/s) | 392147.5 | Total Throughput (Mbit/s) | 383107.3 |
| Packet sent (Bytes) | 46339108 | Packet sent (Bytes) | 1959051 |
| Packet received (Bytes) | 1959120 | Packet received (Bytes) | 46339056 |
| Total retransmits | 22 | Total retransmits | 0 |
| Total errors | 0 | Total errors | 0 |

IV. Conclusions

In this paper, we demonstrated the benefits of RailCat 5e twisted-quad copper cabling for the ETB physical layer region consisting of inter-car and/or inter-consist, advanced rail-approved ETBNs and connectors assemblies for higher data transmission in the harsh railway environment for the targeted application of ETB communication systems with no error and an insignificant retransmission rate. With this new approach, train operating companies can reliably support traffic on-board trains up to the level of 5G and beyond communication systems to enhance passengers’ level of experience and expectations, increase take-up, improve predictive and preventive maintenance leading to reductions in operational expenditure due to downtime. The achievable data rate 40 Gb/s using this method exceeds the current internationally specified standards, referenced in [6] by up to 90 percent. This approach is pushing the frontiers of knowledge, and it is expected that the referenced industry standards will be revised to accommodate higher capacity in the near future.

Future direction of this research would focus on exploring cutting-edge wireless technologies, specifically the IEEE 802.11ax [44] and IEEE 802.11be [45] standards. The objective is to further advance seamless Train-to-Shore connectivity, achieving gigabit throughput-per-area in high-density scenarios while extending the range and bolstering the resilience of backbone to service diversity.

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REFERENCES

Stuart D. Walker was born in Dover, U.K in 1952. He received the B.Sc. (Hons) degree in physics from Manchester University, Manchester, U.K., in 1973, the M.Sc. degree in telecommunications systems and Ph.D. degree in electronics from Essex University, Colchester, U.K., in 1975 and 1981 respectively. From 1981-82, he was a post-doctoral research assistant at Essex University. From 1982-87, he was a research scientist at BT Research Lab, and from 1987-88 he was promoted to Group Leader in Submarine Systems Design. He joined Essex University in 1988 as a Senior Lecturer, and was promoted to Reader in 2002 and to Full Professor in 2004. At Essex University, he manages a laboratory concerned with all aspects of Access Networks: the Access Networks Group (ANG). Prof. Walker been involved in many EPSRC and European Union research projects over the years, recent examples being: e-Photon-I and II, BONE, MUSE-I and II, UROOF, STRONGEST, FIVER, OASE, SODALES, NIRVANA, iCirrus and CHARISMA. Much of the EU project work has involved remote collaboration with colleagues across mainland Europe. The ANG has developed a complete and affordable 8K UHDTV set-up, which uses an in-house designed, state-of-the art, video codec. 4K UHDTV is compressed to just 8 Mbit/s on average (a world-class performance). This was featured in a live 4K webcast of a University of Essex graduation: http://vase.essex.ac.uk/projects/streaming-uhd/index.html. More recently, 8K UHDTV data has been compacted to a typical 100 Mbit/s data-rate using new correlation-based workflows to beat the expected 128 (16 x 8) Mbit/s requirement. These video collaboration technologies use affordable subsystems and are scalable to higher resolutions (e.g. 16K); making the technology future-proof. Live 4K video capture is available in-house whilst 8K cameras are now becoming affordable. Railway communications technology has also featured in the ANG portfolio culminating with the commercial deployment 10 GbE backbones for trains in a knowledge – transfer partnership; an impact cited in the University’s REF 2020 submission. A complete 2 kW sound system (including loudspeakers) has been funded by Essex University to support the ANG’s research into remote collaboration and telepresence.